# Advanced optimal GWO-PID controller for DC motor

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## ABSTRACT

The current work aims to use traditional control algorithms and advanced optimization algorithms that was chosen for its ease of control and the possibility of using it in many industrial applications. By setting the appropriate specifications for the simulation model and after conducting the planned tests that simulate different applications of the motor's work within electrical systems, the results proved to obtain good performance of the motor's work, better response, high accuracy, in addition to the speed. The goal is to design and tune a proportional–integral–derivative (PID) controller by grey wolf optimization (GWO) using transfer function (T.F) for a direct current (DC) motor. To adjust the parameters of the traditional controllers using the optimum advanced, an appropriate mechanism and technology from the advanced optimization techniques were chosen, as the gray wolf technology algorithm was chosen as an optimization technique and integral time absolute error (ITAE) to adjust the parameters of the traditional PID controller.

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### 1. INTRODUCTION

Electric motors are generally used in many systems, including household, military, medical and industrial [1], [2]. Direct current (DC) motors are considered one of the most important electric motors [3]–[5]. They have specifications that make their choice to move many machines within different applications [6]–[8]. It is the most appropriate technically with high performance and spatiality, occupying less place due to its small size [9]–[11]. DC motors have occupied a wide range of application fields due to their high performance. DC motors include conventional DC and brushless DC electric motor (BLDC) [12]–[14].

Conventional control systems handle errors resulting from the work of linear systems, any adjustment process for their parameters according to the linear system is designed [15]–[17]. An algorithm is developed to suit the system that needs a control unit to improve its performance [18]–[20]. In order to improve the performance of the work of linear systems and the shortcomings of traditional performance [21]–[23]. Many solutions were found, including the advanced optimization [24]–[26]. They simulate the hunter and prey and the iteration process as a result of similar cases and experience, including the genetic algorithm (GA) and Particle swarm optimization (PSO) as well as the wolves grey wolf optimization (GWO) and there are other ways [27]–[29].

In the current work, the GWO algorithm was chosen to adjust parameters the traditional proportional-integral-derivative (PID) controller to improve the working performance of the DC motor. To control the motor speed no it is necessary to specify a target function such as integral time absolute error (ITAE). The work of this type of techniques depends on the creation of a random number of the hunting

group and the target function determines the hunting behavior and works to estimate the location of the prey through iterative processes to obtain the optimal location of the prey.

#### 2. DC MOTOR AND SIMULATION MODEL

The electric motor works to produce mechanical energy through the torque arising from the passage of an electric current in the coils of the motor [30]-[32]. First, which in turn acts as an influence in the magnetic circuit in the motor and thus produces a magnetic driving force. Second, which in turn generates a torque to rotate the moving part [33]-[35] which is the rotating part of the electric motor, which produces kinetic energy or mechanical [36]–[38]. The electric motor has special specifications that make the selection process for any application easy [39]-[41]. Such as rotational speed, number of poles, capacity, current, and voltage. according to the manufacturer, and each quantity has a symbol and amount. Table 1 represents the specifications of a DC motor that is intended to perform a computer simulation. To represent it mathematically first and to find the transformation function which is used in building the simulation model. A working algorithm is developed to simulate the operation of the engine first without quantum units. Secondly, the engine is run with the traditional control unit and compared with the previous case. Third, the simulations are carried out with the optimized advanced technique and the comparison with the previous cases. After the simulation, it is possible to work on the appropriate design to control the work of the electric motor by setting an algorithm for better performance, higher speed and accuracy in quality to reach a state close to ideal in performance. In Figure 1 show the simple model for DC motor. In Figure 2 show the block diagram for DC motor at Tf = 0 (no load). In Figure 3 show the block diagram for DC motor with Tf (load).



Figure 1. Simple model for DC motor



Figure 2. Block diagram for DC motor at Tf = 0 (no load)



Figure3. Block diagram for DC Motor with Tf(load)

The electric motor has an electric part and it can be put into a mathematical model with equations through which the theoretical specifications can be obtained and the ideal calculations are calculated through it. Equations (1) and (2) include both voltage and power, respectively. The electric motor has a mechanical second part that can be represented in equations (3) and (4) for speed and torque straight [42]–[44].

$$e_{a} = i_{a}.R_{a} - e_{b}V$$

$$e_{b} = K_{e}.\omega_{m}V$$
(1)

Where,  $K_e = back emf constant$ ,  $\omega_m = motor speed (rad/sec)$ 

$$e_a = (1.2 * 2) + (0.06 * 500) = 32.4 \text{ Volt}$$
  
P =  $\omega_m$ . TV (2)

Where, P =shaft power, T =motor torque

$$P = 500 * 0.108 = 54$$
 watt

Where,  $e_a(t)$  = armature terminal voltage,  $i_a(t)$  = armature current,  $R_a$  = armature resistance,  $L_a$  = armature inductance,  $e_b(t)$  = back emf.

$$\omega_{\rm m} = \frac{K_{\rm T} \cdot e_{\rm a} - ({\rm T} - {\rm T}_{\rm f}) \cdot R_{\rm a}}{K_{\rm T} \cdot K_{\rm e}}$$

$$\omega_{\rm m} = \frac{e_{\rm a} - i_{\rm a} \cdot R_{\rm a}}{K_{\rm e}}$$

$$\omega_{\rm m} = \frac{K_{\rm T} \cdot e_{\rm a} + ({\rm T}_{\rm f}) \cdot R_{\rm a}}{K_{\rm T} \cdot K_{\rm e}}$$

$$\omega_{\rm m} = (0.06 * 32.4) + [(1.2 * 0.012)/(0.06 * 0.06)] = 536 \, rad/s$$
(3)

Where,  $K_T$  = torque constant,  $T_f$  = static friction torque,  $j_m$  = rotational inertia,  $B_m$  = viscous friction.

$$T = K_{T} \cdot i_{a} - T_{f}N - m$$

$$T = (0.06 * 2) - 0.012 = 0.108 N - m$$
(4)

The mathematical model of the electric motor can also be described by the mathematical equations:

Electrical equation: 
$$e_a(t) = R_a \cdot i_a(t) + L \frac{di_t(t)}{dt} + e_b(t)$$
 (5)

Lap. Tra. of electrical equation:  $E_a(s) = L.s.I_a(s) + R_a + I_a(s)$  (6)

$$I_{a}(s) = \frac{1}{L.s.R_{a}} [E_{a}(s) - E_{b}(s)]$$
(7)

Mechanical equation: 
$$T(t) = j_m \frac{d\omega_m(t)}{dt} + B_m \cdot \omega_m(t)$$
 (8)

Lap. Tra. of mechanical equation:  $T(s) = (j_m \cdot s + B_m) \cdot \omega_m(s)$  (9)

$$\omega_{\rm m}({\rm s}) = [1/({\rm j}_{\rm m}.\,{\rm s} + {\rm B}_{\rm m})].\,{\rm T}({\rm s}) \tag{10}$$

Electromechanical relationships: 
$$e_b(t) = K_E \cdot \omega_m(t)$$
 (11) $T(t) = K_T \cdot i_a(t)$  (11)

Lap. Tra. of electromechanical relationships:
$$e_b(s) = K_E \cdot \omega_m(s)$$
 (13)T(s) = K<sub>T</sub> · i<sub>a</sub>(s) (12)

A transfer function for a DC motor as show in equation 13 [43].

$$\frac{\omega_{\rm m}(s)}{E_{\rm a}(s)} = \frac{G(s)}{1 + G(s).{\rm H}(s)} = \frac{K_{\rm T}}{L_{\rm a}j_{\rm m}.s^2 + (R_{\rm a}j_{\rm m} + B_{\rm m}L_{\rm a}).s + (K_{\rm T}.K_{\rm E} + R_{\rm a}.B_{\rm m})}$$
(13)

T. F for DC motor = 
$$\frac{16.13}{0.00333s^2 + 0.201s + 1}$$
 (14)

Table 1. Parameter	rs of DC motor
Parameters of DC motor	Meaning
$\omega_m$	500 rad/sec
$I_a$	2 A
К <sub>е</sub>	0.06 V.s/red
$K_T$	0.06 N.m/A
T <sub>f</sub>	0.012 N-m
R <sub>a</sub>	1.2 ohm
İm	6.2*10-4 N-m-s/rad
B <sub>m</sub>	1*10-4 N-m-s/rad
L	0.02 H

Hierarchical algorithm, as shown in Figure 4, is called wolf algorithm and can be represented by categories that include the leader, who is from the first level to the top of the hierarchy and is called alpha (a). This category is concerned with making decisions such as the spatial environment for living and hunting and can be male or female or both. The second category is a body considered in the second level of the hierarchy. Hierarchical duty helps the first category to make some decisions and is called beta ( $\beta$ ). The third category is lower than its predecessors in terms of hierarchy called delta ( $\delta$ ), where it is considered a category subject to the orders of other groups and has a role to play, which is being a scapegoat. Below the pyramid is the lowest rank category, the fourth, which is the category of sheikhs, which performs the functions of guard and scouts. It is considered dominant over its predecessor, omega ( $\omega$ ), but is subject to both alpha and betas. In Figure 4 show the hierarchy of GWO. Also, the GWO had flow chart that show in Figure 5.



Figure 4. Hierarchy of GWO



Figure 5. Flow chart for GWO

### 3. SIMULATION MODEL FOR DC MOTOR

### 3.1. Simulation model for DC motor with constan speed

The objective are design of PID controller and tuned it by GWO with transfer function (T.F). The T.F of the system in open loop as show in (16).

T. F for DC motor = 
$$\frac{16.13}{0.00333s^2 + 0.201s + 1}$$

In Figures 10-12 the simulation model for DC motor. First the simulation model for DC motor without controller as show in Figure 10. Second the simulation model for DC motor with PID controller as show in Figure 11. Then, simulation model for DC motor with GWO-PID controller as show in Figure 12.

To improve the performance of the work of systems asks for the search for performance indications, which are placed to be indications to improve the performance of systems. To design appropriate controllers for any system are tested through one or more performance indications for a high effective system and within the required specifications. These indications can be recognized through the following equations for each type: i) Integral absolute error (IAE) as show in (15). Also in Figure 6 show block diagram forsimulation model of IAE; ii) Integral square error (ISE) as show in (16). Also in Figure 7 show block diagram for simulation model of ISE; iii) Integral of the product between the squared time and absolute error (ISTAE) as show in (17). Also in Figure 8 show block diagram for simulation model of ISTAE; and iv) ITAE integrates

the absolute error multiplied by the time as show in (18). Also in Figure 9 show block diagram for simulation model of ITAE.

$$IAE = \int |e(t)|dt$$
(15)



Figure 6. Block diagram for simulation model of IAE

$$ISE = \int e^2(t) dt$$

(16)



Figure 7. Block diagram for simulation model of ISE

ISTAE = 
$$\int t^2 |e(t)| dt$$

(17)



Figure 8. Block diagram for simulation model of ISTAE

$$\text{ITAE} = \int t |e(t)| dt$$

(18)



Figure 9. Block diagram for simulation model of ITAE

Figure 10 shows the block diagram for simulation model for DC motor without controller. Figure 11 shows the block diagram for simulation model for DC motor with PID controller. Figure 12 shows the block diagram for simulation model for DC motor with GWO-PID controller.



Figure 10. Block diagram for simulation model for DC motor without controller



Figure 11. Block diagram for simulation model for DC motor with PID controller



Figure 12. Block diagram for simulation model for DC motor with GWO-PID controller

### 3.2. Simulation model for DC motor with variable speed

In this section there are three parts, in Figures 13-15 simulation model for DC motor of variable speed. First, the simulation model for DC motor of variable speed without controller as show in Figure 13. Second, the simulation model for DC motor of variable speed with PID controller as show in Figure 14. Last, the simulation model for DC motor of variable speed with GWO-PID controller as show in Figure 15.



Figures 13. Simulation model for DC motor of variable speed without controller



Figures 14. Simulation model for DC motor of variable speed without controller





### 4. SIMULATION RESULTS OF DC MOTOR

4.1. Simulation results of DC motor with constant speed

This section includes three parts. First by using the simulation model in Figure 10. In this part the simulation results as show in Figure 16. Second by using the simulation model in Figure 11. In this part the simulation results as show in Figure 17. Then, by using the simulation model in DC motor with GWO-PID controller in this part the simulation results as show with comparative among for these simulation pars as show in the Table 2 and in Figure 18.



Figure 16. Simulation results for DC motor without controller

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Figure 17. Simulation results for DC motor with PID controller



Figure 18. Simulation results for DC motor without controller, with PID and GWO-PID controller

Table 2. Simulation results with different algorithm					
Algorithms	Over shoot (%)	Under shoot (%)	Rise time (sec)		
DC motor without Controller	22.840	5.675	0.015		
DC motor with PID Controller	5.851	1.871	0.0021		
DC motor with GWO-pid controller		-	0.002		

#### 4.2. Simulation results of DC motor with variable speed

In this section there are three parts, in Figures 19-21 simulation results for DC motor of variable speed. First the simulation response for DC motor of variable speed without controller as show in Figure 19. Second the simulation response for DC motor of variable speed with PID controller as show in Figure 20. Third the simulation response for DC motor of variable speed with GWO-PID controller as show in Figure 21.



Figure 19. Simulation response for DC motor of variable speed without controller



Figure 20. Simulation response for DC motor of variable speed with PID controller



Figure 21. Simulation response for DC motor of variable speed with GWO-PID controller

In Figure 22 show the ITAE with each iteration. In Figure 23 show the step response. In Figure 24 show the wondow of T.F code for DC motor. In Figure 25 show the wondow of code for GWO-PID.







Figure 23. Step response

1	-	clc;clear all;close all;		
2		%T.F of DC Motor		
3	-	ns=[16.13];		
4	-	ds=[0.00333 0.201 1];		
5	-	G=tf(ns,ds)		
6	-	Gf=feedback(G,1)		
7	-	<pre>step(Gf);</pre>		
8	—	hold on		
Co	Command Window			
New to MATLAB? See resources for Getting Started.				

```
G =

16.13

0.00333 s^2 + 0.201 s + 1

Continuous-time transfer function.

Gf =

16.13

0.00333 s^2 + 0.201 s + 17.13

fg
```



Advanced optimal GWO-PID controller for DC motor (Ghada Adel Aziz)

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Untitled.m × +				
11				
12		%GWO parameters		
13				
14	-	iter=100;		
15	-	pop=60;		
16	-	a_gwo=2;		
17	-	var=3;		
18				
19		%search space		
20	-	al=0; %lower bound		
21	-	au=1000; %upper bound		
22	-	c_cf=0;		
23		<pre>%initialization</pre>		
24	-	for p=1:pop		
25	-	for n=1:var		
26	-	x(p,n) = al + rand*(au - al);		
27	-	end		

Figure 25. Window of code for GWO-PID

#### 5. CONCLUSION

The current study was conducted by simulating the operation of a DC motor with working conditions for three cases that were selected and suggested to verify the preference for using the optimal advance and compare it with the two cases of no control systems. Others with the presence of control using the traditional system, a traditional control unit. The results demonstrated the superiority of the optimum advanced over the traditional progress unit in terms of response speed, time to reach a stable state, accuracy and quality. To adjust the parameters of the traditional controllers using the optimum advanced, an appropriate mechanism and technology from the advanced optimization techniques were chosen, as the Gray wolf technology algorithm was chosen as an optimization technique and ITAE to adjust the parameters of the traditional PID controller. The response speed, high accuracy, stability time and overtaking are higher and lower compared to different cases of working conditions that simulate real time.

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